

# ENDURANCE ANALYSIS OF BATTERY-POWERED MULTICOPTERS: DERIVATION OF PROPER CUT-OFF VOLTAGE USING DC MOTOR MODEL

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#### **Abstract**

This work discusses the hover endurance of the battery-powered multicopters from the viewpoint of the battery cut-off condition. In the existing works about the endurance of battery-powered UAVs, the duration for which the battery can supply the required power for steady flight was assumed as flight time. Therefore, the required power based cut-off was assumed in the existing works. On the other hand, the authors derived a cut-off condition according to the *required voltage* for hovering, using a DC motor equivalent circuit model and a battery discharge property model. The authors show that the required voltage based cut-off condition is tighter than the required power based condition, and the former dominates the endurance if typical ESCs are used. The effective capacity depends on the load states which are classified according to the cut-off voltage and discharge property of a battery. The endurance analysis based on the presented results is shown.

**Keywords:** UAVs, Multicopter, Hover endurance, DC motor equivalent circuit, Discharge cut-off condition

#### 1. Introduction

Nowadays, demands for payload and endurance improvement of multicopters are strongly growing as their application area expands to various industry. To improve the endurance, the use of new power source is attempted (ex. fuel cell[1] or gas-electric system[2]). However the lithium secondary battery powered multicopters are still dominant for the practical use.

The endurance of an electric-powered aircraft basically depends on the required power for steady flight and the effective capacity of a battery. Traub[3, 4, 5] proposed an estimation procedure of the endurance of electric-powered aircraft. In these results, required power for steady level flight is derived from the aerodynamics of the aircraft. An effective capacity of the battery is derived by using Peukert's law[6, 7] (an empirical formula which expresses the reduction in battery capacity for higher discharging rate). The endurance is calculated as the time for which a battery can supply required power. Therefore, a proper estimation of effective capacity is an important issue.

However, the estimates from Peukert's law has some vagueness, because Peukert's law does not depend on a cut-off voltage. Although rated capacity (i.e. the capacity determined from a standard discharge test) is proper to individual batteries, effective capacity varies depending on not only the discharge rate but the cut-off voltage.

One of the purposes of this work is to propose a proper cut-off condition for hovering flight. In [3], the time at which a battery cannot supply the required power is considered as a cut-off condition. However, a voltage based expression of cut-off condition is necessary for estimating effective capacity. Furthermore, typical ESCs (Electric Speed Controllers) do not have voltage step-up function, therefore there are two candidates for the cut-off voltage, namely, required power and require voltage based ones.

In this work, the authors determine the required motor voltage for hovering based on momentum theory and DC motor equivalent circuit model. The authors show that a condition from the required voltage is tighter than another one from the required power, and the former dominates the endurance of the multicopters using standard ESCs. Then, load states in hovering are classified in three cases from the required voltage and D-V property of the battery. The effective capacity is also determined according to the proposed load states. As an application of the above results, evaluation of the hover time and load state of the target multicopter (enRoute PG-560) for several configuration is shown.

# 2. Required motor voltage for hovering

In this paper, a hovering flight of a multicopter in still air is considered. The electric propulsion system consists of fixed-pitch propellers, blushless DC motors with ESCs (Electric Speed Controller) and lithium polymer (LiPo) batteries is assumed. For simplicity, assume that

# Assumption 1.

- 1. Each propeller is identical and not overlapped.
- 2. Each propeller is directly driven by a motor.
- 3. Inductance of a motor is small.

A set of a blushless DC motor with an ESC and a connected propeller is called a "propulsion unit" throughout this paper. An electric propulsion system consists of a LiPo battery<sup>1</sup> and several propulsion units electrically connected in parallel to the battery.

According to the Assumption 1, the loading of one propeller is

$$W_p = \frac{W}{n_r},\tag{1}$$

where W is take-off weight,  $n_r$  is the total number of propellers. Thrust coefficient  $C_T$  and torque coefficient  $C_O$  are defined as [9]

$$C_T := \frac{T_p}{\rho \pi R^2 (\omega R)^2}, \tag{2}$$

$$C_Q := \frac{Q_a}{\rho \pi R^3 (\omega R)^2}, \tag{3}$$

where  $T_p, Q_a$  are thrust and torque,  $\rho$  is air density, R is a propeller radius,  $\omega$  is an angular velocity of a propeller. Reynolds number effect on  $C_T, C_Q$  is small and neglected in this paper. Furthermore,  $C_T, C_Q$  are treated as constant throughout this paper because the advance ratio in hovering flight is fixed at zero.

A balance of thrust and loading of the propeller in hovering is

$$T_p = W_p. (4)$$

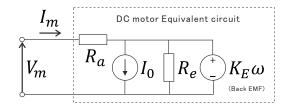
Adopting the propeller induced velocity derived from simple momentum theory, the work done by the propeller on its inflow per unit time is derived as[9]

$$P_{ih} = W_p \sqrt{\frac{W_p}{2\rho \pi R^2}}. (5)$$

This  $P_{ih}$  is the minimum power required for each propulsion unit in hovering. From (2) and (4), angular velocity of a propeller in hovering is derived as

$$\omega_h = \sqrt{\frac{W_p}{C_T \rho \pi R^4}}.$$
(6)

#### **ENDURANCE ANALYSIS OF BATTERY-POWERED MULTICOPTERS**



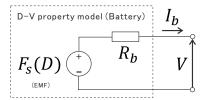


Figure 1 – DC motor equivalent circuit model

Figure 2 - D-V property model

On the other hand, torque of a DC motor is represented as

$$Q_m = K_T (I_m - I_{nl}(\omega)) \tag{7}$$

by using DC motor equivalent circuit model[10] (See Figure 1), where  $K_T$  is torque constant,  $I_m$  is motor current,  $\omega$  is the angular velocity,  $I_{nl}(\omega) := I_0 + (K_E/R_e)\omega$  represents no-load current of a motor. In hovering, a balance of propeller and motor torque,

$$Q_a = Q_m, (8)$$

holds. Therefore the motor current in hovering is derived from (3), (6) and (7) as

$$I_{mh} = \frac{C_Q \rho \pi R^5}{K_T} \omega_h^2 + I_{nl}(\omega_h). \tag{9}$$

The first term of the right hand side of (9) is usually much larger than  $I_{nl}$  for practical multicopter configurations. Therefore, an approximation of  $I_{mh}$  is

$$I_{mh} \simeq \frac{C_Q}{C_T} \frac{R}{K_T} W_p. \tag{10}$$

From the Assumption 1, effect of motor inductance is ignored. Therefore

$$I_m = \frac{V_m - K_E \omega}{R_a} \tag{11}$$

holds. In (11),  $V_m$  is motor voltage,  $K_E$  is back EMF (Electromotive Force) constant,  $R_a$  is the coil resistance of a motor. Note that  $K_T = K_E$  for DC motors.

From (11), the required motor voltage to rotate a propeller at  $\omega_h$  is determined as

$$V_{mh} = R_a I_{mh} + K_E \omega_h \tag{12}$$

Note that  $V_m$  in the above discussion is not equal to the battery-side terminal voltage of an ESC. A typical ESC is an inverter without voltage step-up function. Therefore, the ESC applies PWM (Pulse-Width Modulation) to the voltage input to a motor, and steps down the one-cycle-average (OCA) of the input to desired  $V_m$ . A voltage applied (in the meaning of OCA) to a motor is determined from the voltage supplied to an ESC and throttle position as

$$V_m^{(OCA)} = V^{(ESC)} \cdot f_{th}(T), \tag{13}$$

where  $0 \le T \le 1$  represents normalized throttle position and  $f_{th}$  represents throttle characteristic function. The  $f_{th}$  is any monotonically increasing function such that

$$0 < f_{th}(T) < 1 \text{ for } 0 < T < 1, \tag{14}$$

hence  $f_{th}(0) = 0$  and  $f_{th}(1) = 1$ .

<sup>&</sup>lt;sup>1</sup>A battery is usually made of LiPo cells connected in series/parallel to obtain desired voltage and capacity.

# 3. Cut-off condition and effective capacity

The discharge capacity of a battery is defined as the quantity of the extracted electric charge in constant rate discharge[11]. In what follows, the authors simply use the term "capacity" as the meaning of discharge capacity. The end of discharge is when the battery's terminal voltage drops to some cutoff voltage. Because of battery internal resistance, the higher discharge rate is, the lower terminal voltage is, and it causes the reduction in battery capacity. Therefore, an effective capacity depends on the condition of discharge (i.e. discharge current and cut-off voltage).

IEC 61960-3 standard[11] defines the condition and procedure to determine "rated capacity" of the lithium secondary battery. In the definition, discharge rate is specified as  $0.2C^{-2}$ , and the cut-off voltage is what should be specified by the battery manufacturer (usually about 2.5 [V] per cell).

The terminal voltage during discharge is lower than open circuit voltage (OCV) because of the internal voltage drop caused by battery impedance. The propulsion system in hovering requires constant power. As the terminal voltage gradually drops by discharging, throttle control is required to increase battery current and maintain constant power.

In this work, simplified battery model (let us call "*D-V* property model") only considering internal resistance component (Figure2) is used under the assumption of the rate of change of discharge current is not high in hovering.

$$V(D,I) = F_s(D) - R_b I \tag{15}$$

In (15), V is battery terminal voltage, I is discharge current,  $0 \le D \le 1$  is the depth of discharge (DoD).  $D(t) := I \cdot t/Q_{rat}$  for discharge time t, where  $Q_{rat}$  represents rated capacity. Therefore D = 0 means full charge and D = 1 means full discharge.  $F_s$  is OCV and  $R_b$  is internal resistance. Let us assume that

**Assumption 2.**  $F_s(D)$  is monotonously decreasing and  $F_s(D) > 0$   $(0 \le D \le 1)$ .

Therefore the maximum of the OCV is  $F_s(0) =: V_{full}$ . From (15), an OCV-based cut-off voltage  $V_{end}$  is defined as

$$V_{end} := V_{end}^{rat} + R_b I_{0.2C}, \tag{16}$$

where  $V_{end}^{rat}$  is the terminal voltage based cut-off for the discharge test to determine rated capacity.  $I_{0.2C}$  represents the discharge current which corresponds to 0.2C for each cell. Furthermore, the terminal voltage based cut-off for a discharge current  $I_b$  is defined as  $V_{end}^{TV}(I_b) := V_{end} - R_bI_b$  from (15). For simplicity, let us make an assumption for ESC and battery.

# Assumption 3.

- 1. The current between a battery and an ESC is not changed by the switching operation in the ESC.
- 2. The voltage drop because of the wire resistance between a battery and an ESC is ignored. The power loss by switching operation in the ESC is also ignored.
- 3. The temperature of a battery does not change in hovering. At the start of hovering (or equivalently, discharging), the battery is fully charged and in steady state.

Firstly, consider a hovering condition from the viewpoint of required power. For each propulsion unit, the required electric power in hovering is calculated as  $P_{eh} := V_{mh}I_{mh}$ . A balance of the power consumed by the propulsion system and supplied by the battery is

$$n_r P_{eh} = V(D, I_b)I_b = F_s(D)I_b - R_b I_b^2,$$
 (17)

where  $I_b$  is battery current.

If a multicopter can hover, the equation (17) must have at least one real  $I_b$  as a solution. Therefore the power-based condition for the hovering at the initial time (Remember that D=0 at the initial time) is derived from the discriminant of (17) as

$$F_s(0) \ge 2\sqrt{R_b n_r P_{eh}} = 2\sqrt{R_b I_h \cdot V_{mh}} =: V_{sp},$$
 (18)

<sup>&</sup>lt;sup>2</sup>"C" means "Capacity rate". The capacity rate is defined as the ratio of discharge current [A] to rated capacity [Ah].

where the total of the motor current in hovering is defined as  $I_h := n_r I_{mh}$ . We can assume input impedance of an ESC is higher than  $R_b$ . Therefore

$$I_b(D) = \frac{F_s(D) - \sqrt{F_s^2(D) - V_{sp}^2}}{2R_b}$$
 (19)

from (17).

Secondly, consider another hovering condition from the viewpoint of required voltage. For a fixed  $V_{mh}$ , the minimum voltage should be supplied to the ESC is also  $V_{mh}$ , from (13) and (14) with T=1, then  $I_b=I_h$  from the balance of power.

Therefore, if the multicopter can hover, the initial terminal voltage must exceed the required motor voltage for hovering. From the Assumption 2, a voltage-based condition for the hovering at the initial time is derived as

$$V(0,I_h) > V_{mh} \Leftrightarrow F_s(0) > V_{mh} + R_b I_h =: V_{sh}.$$
 (20)

From the inequality of arithmetic and geometric means and considering  $V_{mh} \gg R_b I_h$  for practical working conditions,

$$V_{sh} > V_{sp} \tag{21}$$

holds. Therefore an essential voltage condition which determines hover duration is (20), rather than (18).

From (6), (10) and (12),  $V_{sh}$  can be described as

$$V_{sh}(W) = \left(\frac{R_a}{n_r} + R_b\right) \cdot \frac{C_Q R}{C_T K_E} \cdot W + K_E \sqrt{\frac{W}{n_r C_T \rho \pi R^4}}.$$
 (22)

Let us call  $V_{sh}$  in (22) "voltage-required" in hovering. We can see  $V_{sh}$  is monotonously increasing for take-off weight W. Note that  $T_h := n_r T_p(\omega_h)$ ,  $C_Q R/C_T K_E = I_h/T_h = I_h/W =: b_{mh}$  is the specific motor current in hovering. From (22),  $V_{sh}$  have a minimum with respect to  $K_E$ .

$$\min_{K_E}(V_{sh}) = 2\left\{\sqrt{\frac{W}{n_r C_T \rho \pi R^4}} \left(\frac{R_a}{n_r} + R_b\right) \cdot \frac{C_Q R}{C_T} \cdot W\right\}^{\frac{1}{2}} = 2\omega_h K_E^{min},\tag{23}$$

for

$$K_E^{min} = \left\{ \sqrt{\frac{n_r C_T \rho \pi R^4}{W}} \left( \frac{R_a}{n_r} + R_b \right) \cdot \frac{C_Q R}{C_T} \cdot W \right\}^{\frac{1}{2}}.$$
 (24)

The inequality (20) implies there exists some  $t_h > 0$  such that  $F_s(D(t)) \ge V_{sh}$  for  $0 \le t \le t_h$ . Therefore, the hover duration is interpreted as the duration for which the  $F_s$  ("voltage-available" in the sense of OCV) exceeds the  $V_{sh}$  (voltage-required).

From the viewpoint of the battery capacity,  $V_{sh}$  in (22) is considered as an OCV based proper cut-off voltage for a certain load state defined in the following. The load states of the propulsion system in hovering are classified in three cases shown in Figure 3.

**Admissible load** ( $V_{end} < V_{sh} < V_{full}$ ) At the end of hovering, T = 1 and  $V = V_{mh}$ . This means there is no room for increasing throttle position to keep hovering. Therefore, the cut-off voltage is  $V_{sh}$  in (22) and the effective capacity is less than  $Q_{rat}$ .

**Rated load (** $V_{sh} \leq V_{end}$ **)** At the end of hovering,  $T \leq 1$  and  $V = V_{end}^{TV}(I_b(1)) (\geq V_{mh})$ . This means the battery is fully discharged until the rated capacity is achieved. Therefore, the cut-off voltage is  $V_{end}$  in (16) and the effective capacity is equal to  $Q_{rat}^3$ .

**Overload** ( $V_{full} \leq V_{sh}$ ) Impossible to hover.

 $<sup>^3</sup>$ Note that the effective capacity does not exceeds  $Q_{rat}$  even if  $V_{sh} < V_{end}$ 

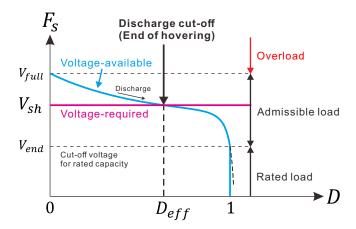


Figure 3 - Load states

From the above discussion, the effective capacity in hovering is described as

$$Q_{eff}(V_{sh}) = D_{eff}(V_{sh})Q_{rat}.$$
(25)

The  $D_{eff}$  is the specific effective capacity of the battery or similar to Coulombic efficiency (CE) in battery analysis. Using the D-V property model in (15),

$$D_{eff} = \begin{cases} F_s^{-1}(V_{sh}) & (V_{end} < V_{sh} < V_{full}) \\ 1 & (V_{sh} \le V_{end}) \end{cases},$$
 (26)

where  $F_s^{-1}$   $(0 \le F_s^{-1} \le 1)$  is the inverse function of  $F_s$ .

#### 4. Hover endurance

Consider discharging in constant current  $I_b$  for dt. The capacity  $dQ = I_b dt$ , and  $t = \int dt = \int (1/I_b) dQ$ . From (25), the hover time  $t_h$  for given  $V_{sh}$  is represented as

$$t_h = \int_0^{D_{eff}} \frac{Q_{rat}}{I_h(D)} dD. \tag{27}$$

From the definition of the admissible and rated load, a lower bound of  $t_h$  is obtained as

$$t_h \ge D_{eff} \cdot \frac{Q_{rat}}{W} \cdot b_{mh}^{-1} \cdot \frac{I_h}{I_b(1)},\tag{28}$$

where  $Q_{rat}/W$  is capacity-weight ratio. Note that  $I_h/I_b(1) \ge 1$  ( $I_h/I_b(1) = 1$  for the admissible load). If we adopt  $(I_b(0) + I_b(D_{eff}))/2$  as a representative battery current, the next approximation is obtained.

$$t_h \simeq D_{eff} \frac{2Q_{rat}}{I_h(0) + I_h(D_{eff})} =: \tilde{t}_h. \tag{29}$$

#### 5. Endurance analysis of multicopter

#### 5.1 Measurement of propulsion system parameters

The target aircraft of the endurance analysis is enRoute PG-560 quadcopter (Figure 4). Parameters of the propulsion system are shown in Table 1.

The authors developed a test bench for the measurement of the characteristics of electric propulsion systems (Figure 5). This equipment can measure the static thrust, torque, rotational speed of the propeller, and the consumed electric power by the propulsion unit.

The normalized throttle positions for the measurement are

$$T \in \{0.2, 0.3, 0.4, 0.5, 0.6, 0.7, 0.8, 0.9\}. \tag{30}$$





Figure 4 – enRoute PG-560 quadcopter

Figure 5 – Test bench for electric propulsion system

The pulsewidth of PWM signal to control an ESC is determined as (1.0+T)[ms]. The rotational speed of the propeller was measured by a Hall sensor (Melexis US1881). Propeller thrust and torque was measured by a 6-axis force torque sensor (WACOH-TECH WEF-6A200-4-RC5). Electric power consumption at the inlet of an ESC is measured by using a current shunt and a power monitor (Texas Instruments INA226). A switching DC power supply (TEXIO PSW-720L30) is used as the power source of the propulsion unit.

For the propulsion system of PG-560, the rotational speed of the propeller is almost in proportion to the throttle position. A least-square fitting model is  $\omega[\text{rad/s}] = 670.77T - 7.1314$  ( $0 \le T \le 1$ ). Measured  $C_T$  and  $C_Q$  are shown in Figure 6.  $P_e$ ,  $P_a - P_e$  (power loss) and  $\eta := P_a/P_e$  (motor efficiency) are shown in Figure 7.  $K_E$ ,  $R_a$  are estimated from no-load rotation tests and shown in Table 2. Calculated KV rating 4 of the motor is 333 [rpm/V] This result is very similar to the value from the specification in Table 1 (330 [rpm/V]).

# 5.2 Modeling of *D-V* property

An OCV property of a battery is modeled by using "Nernst model" [8] in this work. According to this model, the OCV property of a cell is represented as

$$f_{ocv}(D) = E^0 + a\ln(1 - D + \varepsilon_1) + b\ln(D + \varepsilon_2) + c(1 - D + \varepsilon_1)^{-1} + d(1 - D + \varepsilon_1),$$
 (31)

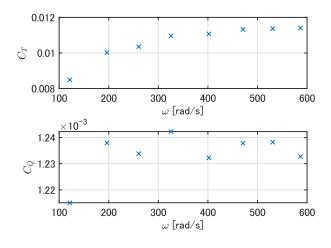
where  $E^0$  is standard electrode potential. The parameters (a,b,c,d) are the characteristics parameters of a cell. The adjustment parameters  $\varepsilon_1, \varepsilon_2 > 0$  are introduced to keep  $f_{ocv}$  finite at D = 0,1. A battery is usually made of several cells connected in series/parallel. Furthermore, batteries may be connected each other on board. Therefore, assume the assembly of all cells on board as a virtual large battery.

## Assumption 4.

Table 1 – Propulsion system specifications (enRoute PG-560)

Propeller (CFRP)	enRoute EXC044	Diameter 15[in] (=381[mm]), Pitch 5.5[in]
Brushless DC motor	enRoute Zion 4631-330KV	KV 330[rpm/V], 24 Poles
ESC	enRoute Zion M40A	max current 40[A], LiPo cells 2~6
Battery (LiPo)	enRoute Zion 99Wh	6 Cells, Weight 0.797[kg]

 $<sup>{}^{4}</sup>KV = (K_{E}[V/rpm])^{-1}.$ 



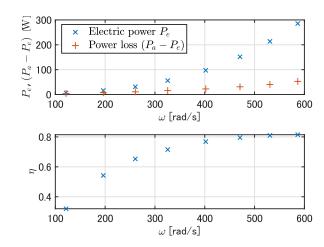


Figure  $6 - C_T$  and  $C_Q$  (enRoute EXC044 propeller)

Figure 7 –  $P_e$ ,  $(P_a - P_e)$  and  $\eta$  (enRoute Zion 4631-330KV motor with Zion M40A ESC)

- 1. Characteristics of all cells are identical.
- 2. The total number of cells is  $n_s \times n_p$ , where  $n_s, n_p$  are the number of cells connected in series or parallel, respectively.

Then the OCV of the virtual battery is described as  $F_s(D) = n_s f_{ocv}(D)$ . The internal resistance  $R_b$  of the virtual battery is described as

$$R_b = \frac{n_s}{n_p} R_{cell},\tag{32}$$

where  $R_{cell}$  is the internal resistance of a cell.

The results of the constant current discharge tests of enRoute Zion 99Wh LiPo battery is shown on Figure 8.  $R_{cell}$  is estimated from the averaged terminal voltage in  $0.4 \le D \le 0.8$  for each discharge test. The parameters in (31) are determined by using "fit" function from MATLAB R2018 Curve Fitting Toolbox. The fitted parameters are shown in Table 3. A determined D-V property model is shown in Figure 9. Note that early responses (i.e.  $D \simeq 0$ ) in discharge were excluded for the parameter fitting (shown in Figure 9 by "x" marker).

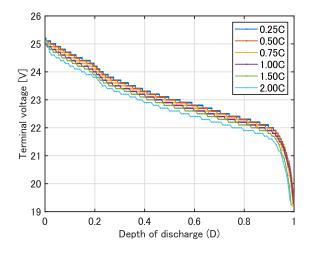
#### 5.3 Endurance analysis

The empty weight<sup>5</sup>  $W_{emp}$  of PG-560 is shown in Table2. A payload is defined as  $W_{pay} := W - W_{emp}$ . In this section, hovering time is numerically evaluated using the measured parameters and D-V property model of the target multicopter. The parameters for the parametric study are take-off weight W

Table 2 – Measured parameters

Body (enRoute PG-560)	$n_r$	4
	$W_{emp}$ [N] (empty weight)	19.6
Propeller (enRoute EXC044)	<i>R</i> [m]	0.19
	$\rho  [kg/m^3]$	1.19
	$C_T$ (Average for $121.6 \le \omega [rad/s] \le 586.7$ )	0.0106
	$C_Q$ (Average for $121.6 \le \omega [rad/s] \le 586.7$ )	0.00123
Brushless DC Motor	$K_E$ [V · s/rad]	0.0287
(enRoute Zion 4631-330KV)	$R_a [\Omega]$	0.20

<sup>&</sup>lt;sup>5</sup>The empty weight is the weight of the multicopter without including batteries and application payloads (ex. a movie camera).



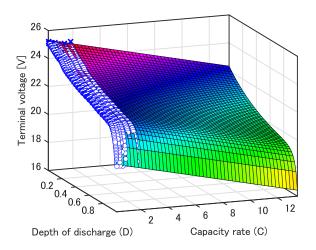


Figure 8 – Terminal voltage in constant current discharge (enRoute Zion 99Wh battery,  $22.5 \pm 1$ [°C])

Figure 9 - D-V property model (enRoute Zion 99Wh battery)

and payload-battery weight ratio  $\lambda_{pay} := W_{batt}/W_{pay}$ . The rated capacity for this study is assumed to be continuous valued, namely, not constrained to the integer multiple of the cell capacity of the test battery ( $Q_{cell}$  in Table2). Throughout the study,  $n_s = 6$  (fixed) and  $n_p = Q_{rat}/Q_{cell}$ . The parameter combination for the study is

$$(W, \lambda_{pay}) \in \{W_{emp} + 2(n-1)[N], n = 1, \dots, 30\} \times \{0.1m, m = 1, \dots, 10\}.$$
 (33)

Figure 10 shows the  $t_h$  in (27). The maximum hover time (36.7 [min]) was obtained for  $(W, \lambda_{pay}) = (5.67 \text{ [kgf]}, 1)$ . Note that the areas colored yellow in Figure 10-12 represent non-practical configurations because of poor weight-thrust ratio  $(T_h^{max}/W < 1.3)$ .

From Figure 11, we can see the large area of the practical configurations corresponds to rated load. The target multicopter is designed for the flight with  $1\sim 2$  batteries (Zion 99Wh). These configurations are represented by two magenta curves in Figure 11. Most part of these lines are included in the rated load area.

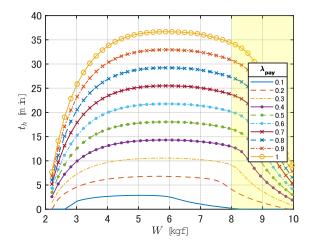
Figure 12 shows the  $\tilde{t}_h$  in (29). Comparing Figure 10 and 12, although an approximation error becomes larger for small W,  $\tilde{t}_h$  is useful to obtain rough estimates. For a large scale parameter study, a piecewise linear approximation of  $F_s^{-1}$  is useful for fast calculation.

#### 6. Conclusions

In this work, the hover endurance of the battery-powered multicopters is discussed from the viewpoint of the battery cut-off condition. Endurance of the electric-powered aircraft is basically depends on required power for steady flight and effective capacity of a battery. In the pioneering works by Traub[3, 4, 5], the use of the discharge time estimated by using Peukert's law[6, 7] was proposed. However, this estimate has a vagueness because Peukert's law does not depend on cut-off voltage which is

Table 3 – Parameters in *D-V* property model (enRoute Zion 99Wh)

$(n_s, n_p)$	(6, 0)
$(\varepsilon_1, \varepsilon_2)$	(0.05, 0.5)
(a,b,c,d)	(-0.2257, -0.6983, -0.0477, -0.0022)
$E^0$ [V]	3.8
$R_{cell}$ [ $\Omega$ ]	0.0083



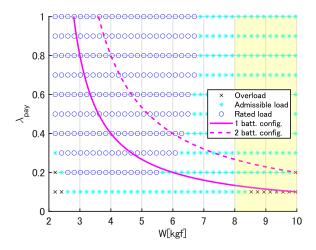


Figure 10 – Hover endurance  $t_h$ 

Figure 11 – Load states for  $(W, \lambda_{pay})$ 

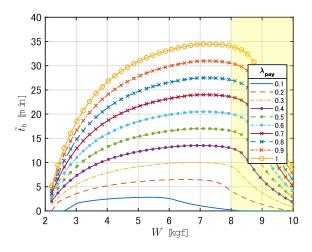


Figure 12 – Approximated hover endurance  $\tilde{t}_h$ 

an essential information to determine the discharge capacity of a battery. Furthermore, the cut-off voltage varies depending on take-off weight.

To elucidate proper cut-off conditions for hovering flight, the authors determined the required motor voltage in hovering, using DC motor equivalent circuit model. After that, the condition from the required voltage in (20) is proved to be tighter than that from the required power in (18) when typical ESCs that do not have voltage step-up function are used.

The hover duration is interpreted as the duration for which the "voltage-available"  $F_s$  exceeds the "voltage-required"  $V_{sh}$ . The authors introduced the classification of load states of the propulsion system by using  $F_s$  and  $V_{sh}$ , and made clear how the effective capacity depends on the load status. Hovering time was defined in a similar manner for engine-powered aircraft, by using specific effective capacity  $D_{eff}$ , rated capacity  $Q_{rat}$  and battery current (or electric charge consumption rate)  $I_b$ .

As an application of the presented results, a parametric study was conducted to evaluate the hover time of the target multicopter (enRoute PG-560) for several combination of the take-off weight and payload-battery weight ratio.

#### **Acknowledgements**

The authors would like to thank M. Kikuchi, technical staff member of Iwate university for developing research equipment. This work is partially supported by the discretionary expense of the dean of the faculty of science and engineering, Iwate university, and Soft-Path Engineering Research Center

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